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3D Electromagnetic Theory of ICRF Multi Port Multi Loop Antenna

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3D ELECTROMAGNETIC THEORY of ICRF MULTI PORT MULTI LOOP ANTENNA

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ABSTRACT

In this report the theory of three dimensional antenna in Ion Cyclotron Resonance Frequency (ICRF) is developed for a plasma with circular magnetic surfaces. The multi loop antenna is located in ITER several ports. Circular plasma and antenna geometry provides new important tools to account for: 1) right loading antenna impedance matrix calculation urgently needed for a matching of RF generator with an antenna; 2) right calculation of an antenna toroidal and poloidal excited spectra because the DIFFRACTION, refraction and REFLECTION effects for the Fast Waves (FW) are in FIRST time are included self consistently in 3D ICRF antenna - plasma treatment; 3) right caculation of RF power deposition profiles because self consistently found 3D antenna - plasma FW excited spectra in

non slab plasma model are important ones in a weakly dissipated plasma for Fast Waves (even for ITER parameters).

In the developed theory multi loop antennae are located in several ITER ports with arbitrary relative toroidal and poloidal positions. It gives great flexibility of investigation possibility for production of an optimal ICRF antenna directivity in conditions of limited toroidal space in a reactor port and a possibility to control widness of excited FW k - spectra to control RF power deposition profiles into a plasma. The above theory allows to calculate RF losses on a Faraday screen bars.

In the model developed each loop is located in a special individual recess in a tokamak port to control in some extent mutual coupling between loops through a vacuum port region and simulteneously to support the Faraday screen bars. The theory developed accounts for two options for poloidal loops: to be located INSIDE of each sub recess (with a smaller mutual loops coupling) and to be located OUTSIDE of each sub recess (more closely to plasma) with an increased antenna loading resistance. Radial antenna loops feeders are taken into account as well.

The MULTI PORT structure of the theory gives another important tool to control mutual coupling of loops: it is possible to locate one port just near another one with, for example, only two loops in each sub PORT just simulating contineous conducting boundary between several loops (besides of sub recess walls). Another possibility is to simulate extended toroidal port size to improve an antenna directivity. The information about plasma properties comes into antenna theory through plasma surface impedance matrix Y_{mm} and is calculated separately by fast MINTOR2 code described in Part A. According to theory developed the ANPORT code has been written. The first runs indicated strong mutual loops coupling of ITER now designed antenna. The work is under progress.

Key Words: Antenna, ICRF, port, multi-loop, current drive, plasma heating, ITER, tokamak, code, plasma, impedance matrix

I. INTRODUCTION. INDEXES and DEFINITIONS

In following theory of multi port, multi loop antennae (each poloidal loop is located in an individual sub-recess) we define meanings of several indexes and clasters of indexes, along with some another termins.

- j number of a port
- 1 "local" number of sub recess in a port
- i-th "global" (over all ports) number of sub loop and a number of EM fields, excited by i-th sub loop.

Numeration of sub loops: since 1st port, from left to right, and from bottom to a top (see Fig. 1ap).

m - poloidal number of a mode in a plasma/SOL,

$$m = m_{min,...,-1,0,1,...,m_{max}}$$

n - toroidal number of a mode in a plasma/SOL

$$n = n_{min}, ..., -1, 0, 1, ..., n_{max}$$

M - poloidal number of a mode in ports/sub recesses

$$M = 0, 1, ..., M_{max}$$

N - toroidal number of a mode in port (regions III, IV),

$$N = 1,2,3,..., N_{max}$$

η - toroidal number of a mode in a sub recess (regions I,II)

$$\eta = 1,2,3,..., \eta_{max}$$

"CLAUSTER" indexes:

 $J = \{ j \mid M \mid \eta \}$ - "global" mode number in a "sub recess"

II. EXPRESSIONS for EM FIELDS in ANTENNA REGIONS I, II

Here are given the expressions for electromagnetic fields in j-th port in sub recess regions I and II (see Fig.1). The components of electrical field, polodal \boldsymbol{E}_{ϕ} and radial \boldsymbol{E}_{r} , satisfying to the boundary conditions on an antenna conducting walls and back plate, are as follows (in l-th

$$E_{\varphi}(r,\varphi,z) = \sum_{i} \sum_{l} \Theta_{l}(\varphi,z) \sum_{M} \cos[\overline{M}(\varphi - \varphi_{j}^{0})] \sin[k_{\eta}(z - z_{jl}^{0})]$$

$$x(-ik_0)\{p_n[A_{ilMm}J_{M}^{'}(p_nr) + B_{ilMm}Y_{M}^{'}(p_nr)] + F_{ilMm}(r)\}$$
 (2.1)

The toroidal wave magnetic field is given by

$$B_{z}(r,\varphi,z) = \sum_{i} \sum_{M,\eta} \cos[\overline{M}(\varphi - \varphi_{j}^{0})] \sin[k_{\eta}(z - z_{ji}^{0})]$$

$$xp_{\eta}^{2} \{A_{jMm} J_{\overline{M}}(p_{\eta}r) + B_{jMm} Y_{\overline{M}}(p_{\eta}r) + F_{jMm}(r)\}$$
(2.2)

The radial electrical fied is:

$$E_{r}(r,\varphi,z) = \sum_{i} \sum_{M,\eta} \sin[\overline{M}(\varphi - \varphi_{j}^{0}) \sin[k_{\eta}(z - z_{j}^{0})]$$

$$x(-\frac{ik_{0}}{r}) \{\overline{M}[A_{jlM\eta_{i}}J_{\overline{M}}(p_{\eta}r) + B_{jlM\eta_{i}}Y_{\overline{M}}(p_{\eta}r)] + G_{jlM\eta_{i}}(r)\}$$
(2.3)

In the above formulas (2.1)-(2.3) J_M and Y_M are the Bessel functions of first kind of fractional \underline{M} order. The function $\Theta_I(\varphi, z)$ is

$$\Theta_l(\phi,z) = \begin{array}{c} 1 & \text{inside of l-th sub recess} \\ \\ 0 & \text{outside of l-th sub recess} \end{array}$$

The coefficients A,B to be found we will supply by upper indexes I,II inside of of sub recess (I) and outside it (II):

$$\dot{A}$$
, B^{I} , $r > s$

$$\dot{A}$$
, B^{II} , $r \le s$

The functions F(r) and G(r) describe the role of radial feeders of current carring loops:

$$F_{jlM\eta i}(r) = \frac{4\pi \tilde{J}_{iM\eta}}{cp_{\eta}^2} \overline{M} u_{M\eta}(r) \theta_s(r) \Delta_{ijl}$$
 (2.4)

$$G_{jlM\eta n}(r) = \frac{4\pi \overline{J}_{iM\eta}}{cp_{\eta}^{2}} (\overline{M}^{2} u_{M\eta}(r) + 1)\theta_{s}(r) \Delta_{ijl}$$
 (2.5)

Here $\underline{J}_{iM_{\eta}}$ is radially independent constant in Fourier component of radial antenna excitation current:

$$J_{r,iM\eta}(r) = \frac{1}{r} \bar{J}_{iM\eta}$$

The function u(r) is a special solution of odinary second order inhomogeneous differential equation and it together with constants $\underline{J}_{iM_{\eta}}$ is decribed in Appendix. The fractional order of Bessels depends on poloidal space extension angle of a port Ψ_p :

$$\overline{M} = \frac{\pi M}{\Psi_n}, \qquad M=0,1,2,...$$

The function $\theta_s(r)$ discriminates regions inside and outside of the sub recess:

$$\theta_s(r) = \{ \\ 0, r \le s \}$$

The subscripted flag Δ_{ijl} discriminates if i-th antenna loop belongs or not to $\{j,l\}$ -th sub recess:

$$\begin{array}{c} \text{ 1, if i-th loop} \in \{\text{j,l}\} \text{ sub recess} \\ \Delta_{\text{ijl}} = \{ \\ \text{ 0, if not} \end{array}$$

The sub recess toroidal wave number is given by

$$k_{\eta} = \frac{\pi \eta}{L_{\tau}}, \quad \eta=1,2,3,...$$

where L_{τ} is toroidal width of a sub recess Another definitions for wave numbers are given as

$${p_{_{\eta}}}^2 \; = \; {k_{_{\! 0}}}^2 - {k_{_{\! \eta}}}^2, \qquad \quad {k_{_{\! 0}}} = \omega \; /c \; , \quad {p_{_{\! \eta}}} = \sqrt{|{p_{_{\! \eta}}}^2|}$$

Here ω is a generator frequency, and c is speed of light.

The boundary condition at the back plate in each recess

$$E(\mathbf{r} = \mathbf{r}_{\mathbf{w}}) = 0$$

and jump of a magnetic field across antenna's poloidal loop if $s > r_s$

$$\left[B_z^{II} - B_z^{I}\right]_{r=s} = \frac{4\pi}{c}j_{\varphi}$$

together with continuity of poloidal electric field at a loop

$$\left[E_{\varphi}^{II} - E_{\varphi}^{I}\right]_{r=s} = o$$

result in system of linear equations:

$$A_{J_i}^{II} = \gamma_{M\eta}^I B_{J_i}^{II} + Q_{J_i}^{II}$$

$$A_{J_l}^I = \gamma_{Mn}^I B_{J_l}^I + Q_{J_l}^I$$

$$A_{J_1}^{II} = A_{J_1}^I + q_{J_1}^I$$

$$B_{Ji}^{II}=B_{Ji}^I+q_{Ji}^{II}$$

where for brevity we used collective index $J = \{ilM\eta\}$. Here

$$\gamma_{Mn_{1}}^{I} = -\frac{Y_{\overline{M}}(p_{\eta}r_{w})}{J_{\overline{M}}(p_{\eta}r_{w})}, \qquad Q_{J_{1}}^{I} = -\frac{F_{J_{1}}(r_{w})}{p_{\eta}J_{\overline{M}}(p_{\eta}r_{w})}$$

$$Q_{Ji}^{II} = Q_{Ji}^{I} + q_{Ji}^{I} - \gamma_{M\eta}^{I} q_{Ji}^{II}$$

$$q_{J_1}^I = w_{\eta} p_{\eta} s [R_1 \dot{Y_{\overline{M}}}(p_{\eta} s) - R_2 Y_{\overline{M}}(p_{\eta} s)]$$

$$q_{J_1}^{II} = -w_{\eta} p_{\eta} s[R_1 J_{\overline{M}}(p_{\eta} s) - R_2 J_{\overline{M}}(p_{\eta} s)]$$

$$R_1 = \frac{4\pi}{cp_{\eta}^2} J_{iM\eta}^{\varphi} + F_{Ji}(s) , \qquad R_2 = \frac{F_{Ji}'(s)}{p_{\eta}}$$

(w_{η} is given in Appendix).

If $i \notin \{j,l\}$ or $i \in \{j,l\}$, but $s < r_s$, then $\{q^I,q^{II}\} = 0$.

III. FIELDS in REGIONS III, IV

In port regions outside of sub recesses area, $b \le r \le r_s$, the fields in any tokamak port, satisfying to boundary conditions, have the following expressions:

$$E_{\varphi}(r,\varphi,z) = \sum_{i} \sum_{j} \Theta_{j}(\varphi,z) \sum_{M,N} \cos[\overline{M}(\varphi - \varphi_{j}^{0})] \sin[k_{N}(z - z_{j}^{0})]$$

$$x(-k_{0}) \{p_{N}[A_{MN_{i}}J_{\overline{M}}(p_{N}r) + B_{MN_{i}}Y_{\overline{M}}(p_{N}r)] + F_{MN_{i}}(r)\}$$
(3)

The toroidal wave magnetic field component in j-th port is:

$$B_{z}(r,\varphi,z) = \sum_{i} \sum_{M,N} \cos[\overline{M}(\varphi - \varphi_{j}^{0})] \sin[k_{N}(z - z_{j}^{0})]$$

$$xp_{N}^{2} \{A_{iMNi}J_{\overline{M}}(p_{N}r) + B_{iMNi}(p_{N}r) + F_{iMNi}(r)\}$$
(3.2)

Radial electrical wave field in that region is as

$$\begin{split} E_r(r,\varphi,z) &= \sum_i \sum_{MN} \sin[\overline{M}(\varphi-\varphi_j^0)] \sin[k_N(z-z_j^0)] \\ & \times (-\frac{ik_0}{r}) \{\overline{M}[A_{jMNi}J_{\overline{M}}(p_Nr) + b_{jMNi}Y_{\overline{M}}(p_Nr)] + G_{jMNi}(r)\} \end{split}$$

The function $\Theta_j(\varphi, z)$ keeps the poloidal electrical field to be zero outside of j-th port:

1 inside of j-th port

$$\Theta_i(\varphi,z) = \{$$

0 outside of it

The coefficients A,B have the indexes to discriminate regions outside of the loop (IV) and inside of the loop if last one is moved otside of sub recess more closely to the plasma (see Fig.1):

$$AB = \{ A^{II}, B^{III}, r \ge s \}$$

$$A^{IV}, B^{IV}, r \le s \}$$

In above formulas the toroidal wave number in a port (outside of sub recesses) is defined as

$$k_N = \frac{\pi N}{L_p},$$
 $N = 1,2,3,...$

where L_p is a toroidal broadness of a port. The functions F(r) and G(r) are given by

$$F_{jMNi}(r) = \frac{4\pi \overline{J}_{iMN}}{cp_{N}^{2}} \overline{M} u_{MN}(r) \theta_{s}(r) \Delta_{y}$$

$$G_{jMN_t}(r) = \frac{4\pi \overline{J}_{MN}}{cp_N^2} (\overline{M}^2 u_{MN}(r) + 1)\theta_s(r)\Delta_y$$

$$\Delta_{ij} = \{ \\ 0 & i \notin \text{ to } j\text{-th port}$$

In the case of poloidal loops being protruded outside of sub recess region, i.e. $s < r_{s_s}$ we account for jump of wave magnetic field across of a loop and continuity of poloidal electric field:

$$\left[B_z^{IV} - B_z^{III}\right]_{r=s} = \frac{4\pi}{c} j_{\varphi} , \text{ if } s \le r_s$$

$$\left[E_{\varphi}^{IV} - E_{\varphi}^{III}\right]_{r=s} = 0$$

These two last conditions create the new linear equations:

$$A_{iMNi}^{IV} = A_{iMNi}^{III} + q_{jMNi}^{III}$$

$$B_{jMNi}^{IV} = B_{jMNi}^{III} + q_{jMNi}^{IV}$$

where

$$q_{jMNi}^{III} = w_N p_N s\{\tilde{R}_1 Y_{\overline{M}}(p_N r) - \tilde{R}_2 Y_{\overline{M}}(p_N r)\}$$

$$q_{_{IMN_{l}}}^{N}=-w_{_{N}}p_{_{N}}s\{\tilde{R}_{1}J_{\overline{M}}^{'}(p_{_{N}}s)-\tilde{R}_{2}J_{\overline{M}}(p_{_{N}}s)\}$$

If $i \notin j$ or $s > r_s$ then q^{III} , $q^{IV} = 0$.

IV. DIVERTOR REGION (region V)

In this section we consider region between bulk plasma (r=a) and a first wall (r=b) and name that as divertor region (region V). In this region poloidal electric field excited by all loops (current global index iz "I") over all ports we expand in Fourier series over plasma toroidal and poloidal harmonics (m,n):

$$\begin{split} E_{\varphi}(r,\varphi,z) &= \sum_{i} \sum_{m,n} e^{im\varphi + ik_{n}z} (-ik_{0}p_{n}) \{A_{mm}^{V} J_{m}^{'}(p_{n}r) + B_{mni}^{V} Y_{m}^{'}(p_{n}r)\} \\ B_{z}(r,\varphi,z) &= \sum_{i} m \sum_{mn} e^{im\varphi + inz} p_{n}^{2} \{A_{mni}^{V} J_{m}(p_{n}r) + B_{mn}^{V} Y_{m}(p_{n}r)\} \end{split} \tag{1}$$

Here toroidal plasma wave number is given as

$$k_n = \frac{n}{R}$$
, n=0,±1,±2,....

$$p_n^2 = k_0^2 - k_n^2$$
, $p_n = \sqrt{|p_n|^2}$

V. BOUNDARY CONDITIONS AT PLASMA SURFACE

At plasma vacuum boundary we require the continuity of polidal electric field and the continuity of a toroidal component of wave magnetic field:

$$E_{\varphi}^{\text{plasma}} = E_{\varphi}^{\text{vacuum}}, \quad B_{z}^{\text{plasma}} = B_{z}^{\text{vacuum}}$$
 (51)

In our previos report [1] we developed the procedure for calculation of wave fields in inhomogeneous plasma. For our antenna electromagnetic charecteristics (antenna impedance matrix, spectrum of excited waves, Ohmic losses on the Faraday screen, etc.) to be found it will be sufficient to make use the plasma surface impedance [1]:

$$Y_{mn} = \left[\frac{E_{qmn}}{B_{zmn}}\right]_{r=q=0} \tag{52}$$

Then from (5.1) and (5.2) one has

$$A_{mn}^{V} = \gamma_{mn}^{V} B_{mn}^{V} , \qquad \tilde{Y}_{mn} = i \frac{p_{n}^{2}}{k_{0} p_{n}} Y_{mn}$$

$$\gamma_{mn}^{V} = -\frac{\bar{Y}_{mn} Y_{m} (p_{n} a) - \dot{Y}_{m} (p_{n} a)}{\bar{Y}_{mn} J_{m} (p_{n} a) - \dot{J}_{m} (p_{n} a)}$$
(53)

From formulas (5.3) we see that coefficient γ^V is closely related with reflection coefficient from plasma.

VI. FIELDS MATCHING AT ANTENNA MOUTH (r = b)

Now we start important procedure of fields matching at an antenna (port) mouth. We stress that in many ITER related 2D full wave codes the fields are matched term by term in Fourier series, so speaking about port antenna but dealing with an antenna located in SOL region. In presented theory we are doing real in port antenna solution by matching TOTAL em fields at an antenna mouth. We equate total fields at r = b, multiply equality by $\exp(-im\phi-ik_nz)$ and integrate first wall surface. Schematically it looks like:

$$\int_{-\pi}^{\pi} d\varphi \int_{-\pi R}^{\pi R} dz e^{-im\varphi - ik_n z} [E_{\varphi}^{V} = E_{\varphi}^{IV}]_{r=b}$$

The same we are doing for the magnetic fields B_z, integrating over j-th port:

$$\int \int_{OVER-J-TH-PORT} d\varphi dz \cos[\overline{M}(\varphi-\varphi_{j}^{0})] \sin[k_{N}(z-z_{j}^{0})] \left[B_{z}^{V}=B_{z}^{IV}\right]_{r=b}$$

This results in important system of linear equations:

$$\sum_{I} \{G_{II}, A_{Ii}^{III} + H_{II}, B_{Ii}^{III}\} = D_{Ii}$$
(6.1)

where

$$D_{li} = q_{li}^{III} J_{\overline{M}}(p_N b) + q_{li}^{IV} Y_{\overline{M}}(p_N b) -$$

$$-\sum_{l}T_{ll'}\{q_{li}^{III}\dot{J_{M}}(p_{N'}b)+q_{l'i}^{IV}\dot{Y_{M}}(p_{N'}b)\},\,$$

The coefficients G and H are given by

$$G_{n'} = T_{n'} J_{\overline{M}}(p_{N'}b) - \delta_{n'} J_{\overline{M}}(p_{N}b)$$

$$H_{n'} = T_{n'} \dot{Y_{\overline{M}}} (p_{n'} b) - \delta_{n'} Y_{\overline{M}} (p_{n} b)$$

As above, the prime at the head of Bessels $\{J,Y\}$ means differentiaton over its arguments, and δ_{ii} is Kroneker function.

In the above integration we made use an orthogonality of trigonometric functions. The real information about plasma, chamber and ports sizes are comes through T coefficient:

$$T_{II'} = \frac{p_{N'}}{\pi^2 R \Delta_M L_p \Psi_p p_N^2} \sum_{mn} R_{mn} \alpha_{jMm} \alpha_{jMm}^* \beta_{jNn} \beta_{jN'n}^*$$

$$R_{mn} = \frac{p_n^2}{p_n} \cdot \frac{\gamma_{mn}^V J_m(p_n b) + Y_m(p_n b)}{\gamma_{mn}^V J_m(p_n b) + Y_m(p_n b)}$$

$$\Delta_{M} = \{ \\ \Delta_{M} = 0 \\ 1, \quad M \neq 0$$

The reflective wave amplitude at an antenna/port mouth is given by

$$B_{mni}^{V} = \frac{1}{4\pi^{2} R p_{n} [\gamma_{mn}^{V} J_{m}(p_{n}b) + Y_{m}(p_{n}b)]}$$

$$x \sum_{l,M,N} \alpha_{jMm}^{*} \beta_{jNn}^{*} p_{N} [A_{jMNi}^{IV} J_{\overline{M}}(p_{N}b) + B_{jMNi}^{IV} Y_{\overline{M}}(p_{N}b)]$$

The integration geometrical coefficients α_{jMm} , β_{jNn} are given in Appendix.

VII. FIELDS MATCHING at SUB RECESSES MOUTHES

Inside of a j-th port we must also to match em fields at sub recesses openings at $r = r_s$. We again multiply the equalities of total em fields by trigonometric functions and integrate in toroidal direction over a port size L_p and over a sub recess size L_r to make use orthogonality of trgonometric functions:

$$\int_{Z_{j}^{0}}^{Z_{j}^{0}+L_{p}} dz \sin[k_{N}(z-z_{j}^{0})] \left[E_{\varphi}^{II}=E_{\varphi}^{III}\right]_{r=s}$$

$$\int_{Z_{s}^{0}}^{d} dz \sin[k_{\eta}(z-z_{jl}^{0})] \left[B_{z}^{II} = B_{z}^{III}\right]_{r=s}$$

Making use the expressions for the fields from previous sections one gets a continuation of linear equations for unknown coefficients for the em fields:

$$\sum_{N} (V_{MNN'} A_{jMN'i}^{III} + W_{MNN'} B_{jMN'i}^{III}) = U_{jMNi}$$
 (7.1)

Неге

$$V_{MNN'} = T_{MNN'}J_{\overline{M}}(p_{N'}r_{s}) - \frac{L_{p}}{2}p_{N}J_{\overline{M}}(p_{N}r_{s})\delta_{NN'}$$

$$W_{MNN} = T_{MNN} Y_{\overline{M}}(p_{N'}r_s) - \frac{L_p}{2} p_{N} Y_{\overline{M}}(p_{N}r_s) \delta_{NN'}$$

The right hand side of Eq. (7.1) includes also a contribution of radial antenna feeders (through the function F):

$$\begin{split} U_{jMNt} &= \frac{L_{p}}{2} \dot{F_{jMNt}}(r_{s}) - \sum_{N} T_{MNN} \dot{F_{jMNt}}(r_{s}) + \\ &\sum_{l,\eta} \mu_{lN\eta} \{ p_{\eta}^{2} R_{M\eta} [F_{jlM\eta t}(r_{s}) + Q_{jlM\eta t}^{ll} J_{\overline{M}}(p_{\eta} r_{s})] - p_{\eta} Q_{jlM\eta t}^{ll} J_{\overline{M}}(p_{\eta} r_{s}) - F_{jlM\eta t}(r_{s}) \} \end{split}$$

Matrix T is given by

$$T_{MNN'} = \frac{2}{L_r} p_{N'}^2 \sum_{l} \sum_{n} (\mu_{lN\eta} \mu_{lN\eta} R_{M\eta})$$

where

$$R_{M\eta} = \frac{p_{\eta}}{p_{\eta}^{2}} \cdot \frac{\gamma_{M\eta}^{I} J_{\overline{M}}^{'}(p_{\eta} r_{s}) + Y_{\overline{M}}^{'}(p_{\eta} r_{s})}{\gamma_{M\eta}^{I} J_{\overline{M}}(p_{\eta} r_{s}) + Y_{\overline{M}}(p_{\eta} r_{s})}.$$

The coefficients $\mu_{lN_{\eta}}$ are given in Appendix. The coefficient B^{II} is expressed as:

$$B_{J_{1}}^{II} = \frac{1}{\gamma_{M\eta}^{I} J_{\overline{M}}(p_{\eta} r_{s}) + Y_{\overline{M}}(p_{\eta} r_{s})} \{ -F_{J_{1}}(r_{s}) - Q_{J_{1}}^{II} J_{\overline{M}}(p_{\eta} r_{s}) + \frac{2}{L_{r} p_{\eta}^{2}} \sum_{N} \mu_{IN\eta} p_{\eta}^{2} [A_{I_{1}}^{III} J_{\overline{M}}(p_{N} r_{s}) + B_{I_{1}}^{III} Y_{\overline{M}}(p_{N} r_{s}) + F_{I_{1}}(r_{s}) \}$$

This completes preparatory work for solving antenna problem on a computer.

VIII. LINEAR SYSTEM OF EQUATIONS

The derived linear equations (6.1) and (7.1) we combine in a compact system For the brevity we use "clauster" index $I = \{j, M, N\}$.

$$\sum_{l} P_{ll'} B_{l'i}^{lll} = \tilde{D}_{li} - \tilde{U}_{li}$$
(8.1)

It means that

$$B_h^{III} = \sum_{i} P_{ii}^{-1} (\tilde{D}_{f_i} - \tilde{U}_{f_i})$$
 (8.2)

where

$$\begin{split} P_{\mathit{II'}} &= \tilde{H}_{\mathit{II'}} - \tilde{W}_{\mathit{MNN'}} \, \delta_{\mathit{IJ'}} \delta_{\mathit{MM'}} \\ \tilde{H}_{\mathit{II'}} &= \sum_{\mathit{I}} G_{\mathit{II'}}^{-1} H_{\mathit{I'I'}} \\ \tilde{D}_{\mathit{I}\iota} &= \sum_{\mathit{I}} G_{\mathit{II'}}^{-1} D_{\mathit{I}\iota} \\ \\ \tilde{W}_{\mathit{MNN'}} &= \sum_{\mathit{N}} V_{\mathit{MNN'}}^{-1} \, W_{\mathit{MN'N'}} \\ \\ \tilde{U}_{\mathit{JMN\iota}} &= \sum_{\mathit{N}} V_{\mathit{MNN'}}^{-1} U_{\mathit{JMN'\iota}} \end{split}$$

For example, $V_{MNN'}^{-1}$ is inverted over indexes N,N matrix $V_{MNN'}$. Then we find

$$A_{jMNi}^{III} = \tilde{U}_{jMNi} - \sum_{N} \tilde{W}_{MNN'} B_{jMN'i}^{III}$$
 (8.3)

It would be possible to make a solution of the linear system "ideologically simple", but with greater requirement on "memory".

IX. TOTAL ANTENNA IMPEDANCE and IMPEDANCE MATRIX

We define total antenna matrix impedance Z through a complex power:

$$P_{RF} = -\frac{1}{2} \int_{V} \vec{j}^{*}(\vec{r}) \vec{E}(\vec{r}) dV = \frac{1}{2} Z I_{0}^{2}$$
 (9.1)

Here I_0 is amplitude of an exciting current in sub loop. (In above definition the multiplyer, responsible for averaging of squared current over loop length has not been accounted for simplicity).

The antenna impedance matrix we define as

$$Z_{ii} = -\int_{V} \vec{j}_{i}^{*}(\vec{r}) \vec{E}_{i}(\vec{r}) dV$$
(9.2)

assuming unit loop current: $I_0 = 1$ SGSM. In Eqs. (9.1), (9.2) integration is performed over volume of excited antenna currents (over the loops). The contribution of radial antenna feeders in an impedance is evaluated by trapezia rules (we recall that "i" is a global sub loop number over all ports and recesses).

In following the indexes j,l (j is port number, l is "local" number of a sub recess in a port) are those, that $\Delta_{i,j} = 1$.

9.1 Loops are deeply in sub recesses

In this case $s > r_s$. Then performing an integration we find:

$$Z_{ii} = \frac{i}{4} k_0 \Psi_p L_r \{ s \sum_1 + \frac{r_w - s}{2} \sum_2 \}$$
 (9.1.1)

Here

$$\sum_{1} = \sum_{Mn} J_{\varphi iM\eta}^{*} \Delta_{M} \{ p_{\eta} [A_{jlM\eta}^{I} J_{\overline{M}}^{'} (p_{\eta} s) + B_{jlM\eta}^{I} Y_{\overline{M}}^{'} (p_{\eta} s)] + F_{jlM\eta}^{I} (s) \}$$
 (9.1.2)

$$\sum_{2} = \sum_{M\eta} \bar{J}_{iM\eta}^{*} \{ \overline{M} [A_{jlM\eta^{i}}^{I} (\frac{J_{\overline{M}}(p_{\eta}s)}{s} + \frac{J_{\overline{M}}(p_{\eta}r_{w})}{r_{w}}) + B_{jlM\eta^{i}}^{I} (\frac{Y_{\overline{M}}(p_{\eta}s)}{s} + \frac{Y_{\overline{M}}(p_{\eta}r_{w})}{r_{w}})] + \frac{G_{jlM\eta^{i}}(s)}{s} + \frac{G_{jlM\eta^{i}}(r_{w})}{r_{w}} \}.$$

$$(9.1.3)$$

The above impedance formulas are useful in many respects. For example from (9.1.1) follows that impedances are proportional to poloidal antenna extent Ψ_p .

In this case antenna sub loops are moved more closely to plasma to increase an active impedance (loading resistance): $s \le r_w$. Direct evaluation of (9.2) gives the impedance matrix:

$$Z_{a'} = \frac{i}{4} k_0 \Psi_p \{ L_p(s \sum_1 + \frac{r_s - s}{2} \sum_2) + L_p \frac{r_w - r_e}{2} \sum_3 \}$$
 (9.2.1)

The contribution of poloidal RF current to the impedance is given by

$$\sum_{1} = \sum_{MN} J_{\varphi_{1}MN}^{*} \Delta_{M} \{ p_{N} [A_{jMN_{1}}^{III} J_{\overline{M}}^{'} (p_{n}s) + B_{jMN_{1}}^{III} Y_{\overline{M}}^{'} (p_{N}s)] + F_{jMN_{1}}^{'}(s) \}$$
(9.2.2)

Contribution of radial feeders outside of a recess is:

$$\sum_{2} = \sum_{MN} \tilde{J}_{iMN}^{*} \{ \overline{M} [A_{jMNi}^{III}, (\frac{J_{\overline{M}}(p_{N}s)}{s} + \frac{J_{\overline{M}}(p_{N}r_{s})}{r_{w}}) + B_{jMNi}^{III}, (\frac{Y_{\overline{M}}(p_{N}s)}{s} + \frac{Y_{\overline{M}}(p_{N}r_{s})}{r_{s}})] + \frac{G_{jMNi}(s)}{s} + \frac{G_{jMNi}(r_{s})}{r} \}$$

$$(92.3)$$

Contribution of part radial feeders, located inside of a recess,to the impedance matrix is given by:

$$\sum_{3} = \sum_{M\eta r} \tilde{J}_{iM\eta}^{*} \{ \overline{M} [A_{jlM\eta r}^{I} (\frac{J_{\overline{M}}(p_{\eta} r_{s})}{r_{s}} + \frac{J_{\overline{M}}(p_{\eta} r_{w})}{r_{w}}) + B_{jlM\eta r}^{I} (\frac{Y_{\overline{M}}(p_{\eta} r_{s})}{r_{s}} + \frac{Y_{\overline{M}}(p_{\eta} r_{s})}{r_{w}})] + \frac{G_{jlM\eta r}(r_{s})}{r_{s}} + \frac{G_{jlM\eta r}(r_{w})}{r_{w}} \}$$

$$(9.2.4)$$

X. RADIAL POWER FLUX

Making use orthogonality eigen functions the radial power flux is expressed in a compact form:

$$\Pi_{\tau} = -\frac{c}{8\pi} \int_{-\pi}^{\pi} d\varphi \int_{-\pi R}^{\pi R} r dz \operatorname{Re}[E_{\varphi} B_{z}^{*}] = \sum_{mn} \Pi_{mn}$$

where at a port mouth r = b each Fourier term is given by

$$\Pi_{mn} = \frac{1}{2w_n} \omega \pi R p_n^2 |B_{mn}^V|^2 \operatorname{Im}(\gamma_{mn}^V)$$
 (10.1)

Here magnetic field \mathbf{B}^{v}_{mn} is a contribution of all loops of multi port antenna

$$B_{mn}^V = \sum_i B_{mni}^V$$

APPENDIX

1a. BESSEL FUNCTIONS

In this report many times appearing Bessel functions J and Y are understood as:

$$J_{\alpha}(pr), \ Y_{\alpha}(pr) \ , \qquad p^{2} \geq 0$$

$$J_{\alpha}(pr), \ Y_{\alpha}(pr) \ = \ \{$$

$$J_{\alpha}(pr), \ K_{\alpha}(pr), \qquad p^{2} < 0$$

where $I_{\alpha}(z)$ and $K_{\alpha}(z)$ are modifyed Bessel functions The Wronskian of Bessels has a form:

$$J_{\alpha}(pr)\dot{Y_{\alpha}}(pr) - \dot{J_{\alpha}}(pr)Y(pr) = \frac{1}{w_{p}pr}$$

where

$$w_p = \{ \\ -1 , p^2 \le 0 \\ p^2 \le 0 .$$

2a. COEFFICIENTS α_{jMm} , β_{jNn} , $\mu_{lN\eta}$

The coefficient alfa is the integral:

$$\alpha_{_{JMm}} = \int_{\varphi_{_{J}}^{0}}^{\varphi_{_{J}}^{0}+\Psi_{_{P}}} \cos[\overline{M}_{_{M}}(\varphi-\varphi_{_{J}}^{0})]e^{\imath m\varphi}d\varphi =$$

$$\begin{cases} -ie^{im\varphi_{j}^{0}}[(-1)^{M}e^{im\Psi_{p}}-1]\frac{m}{m^{2}-\overline{M}_{M}^{2}}, \dots \mid m \mid \neq \mid \overline{M}_{M} \mid \\ = \begin{cases} \frac{\Psi_{p}}{2}e^{im\varphi_{j}^{0}}, \dots \mid m \mid = \mid \overline{M}_{M} \mid \neq 0 \\ |\Psi_{p}, \dots m = M = 0 \end{cases}$$

The coefficient beta is the integral over toroidal wide of a port:

$$\beta_{jNn} = \int_{Z_{j}^{0}}^{Z_{j}^{0} + L_{p}} \sin[k_{N}(z - Z_{j}^{0})] e^{ik_{n}Z} dz =$$

$$\begin{cases} e^{ik_{n}Z_{j}^{0}} [(-1)^{N} e^{ik_{n}L_{p}} - 1] \frac{k_{N}}{k_{n}^{2} - k_{N}^{2}}, \dots \mid k_{n} \mid \neq \mid k_{N} \mid \\ \frac{i}{2} e^{ik_{n}Z_{j}^{0}}, \dots, k_{n} = k_{N} \end{cases}$$

$$\begin{vmatrix} i \frac{L_{p}}{2} e^{ik_{n}Z_{j}^{0}}, \dots, k_{n} = -k_{N} \end{vmatrix}$$

The coefficient mu is integral over toroidal width of l-th sub recess of j-th port:

$$\begin{split} \mu_{lN\eta} &= \int_{Z_{j}^{0}}^{s} \sin[k_{N}(z-z_{j}^{0})] \sin[k_{\eta}(z-z_{j}^{0})] dz = \\ &= \begin{cases} \frac{k_{\eta}}{k_{N}^{2} - k_{\eta}^{2}} \{(-1)^{N} \sin[k_{N}(z_{l}^{0} + L_{r})] - \sin[k_{N}z_{l}^{0}]\}, ...k_{N} \neq k_{\eta} \\ \frac{L_{r}}{2} \cos(k_{N}z_{l}^{0}),, k_{N} &= k_{\eta} \end{cases} \end{split}$$

where $z_i^0 = l(L_r + d_s)$ and d_s is thickness of a sub recess wall in toroidal direction.

3a. FUNCTION U(r)

Presence of radial antenna feeders requires of solution of inhomogeneous differential equation. In cylindrical antenna geometry it is a particular solution of the equation:

$$r^2u''(r) + ru'(r) + (p^2r^2 - \overline{M}^2) = 1$$

where prime means differentiation over r We find this solution numerically by an expansion u(r) in Teylor series over t = r/s-1 argument near of t = 0.

4a. FOURIER COMPONENTS OF SUB LOOP CURRENTS

On Fig. 1ap we scetch antenna mouth how it is seen from a plasma center. In each sub recess are located 2 sub loops (or even one sub loop as an option). The poloidal angle antenna loop and port angle sizes are shown together with coordinates of loop centers in polodal and toroidal directions.

Two operational modes of RF feed are also shown out phase and sin phase feeds, creating different Fast wave exciting spectra.

For the out phase feed the expressions for Fourier harmonics of excing currents are:

$$J_{q : M\eta} = \frac{e^{i\Phi_i}}{D} J_{q \in M} J_{\eta}$$

It is a current in sub pecess. The poloidal current Fourier component in a port region outside of a recess is given by

$$J_{\varphi iMN} = \frac{e^{iu\Phi_i}}{D} J_{\varphi iM} J_{iN}$$

In theses formulas Φ_i and D are respectively phase of i-th loop and toroidal wide of a loop.

The "hatted" currents in main text are

$$\tilde{J}_{iM\eta} = \frac{e^{i\Phi_{i}}}{D} \tilde{J}_{iM} J_{\eta}$$

$$\tilde{J}_{vMN} = \frac{e^{i\Phi_i}}{D} \tilde{J}_{iM} J_{iN}$$

Here

$$J_{\eta} = \frac{4}{\pi\eta} \sin(\frac{\pi\eta}{2}) \sin(k_{\eta} \frac{D}{2})$$

$$J_{iN} = \frac{4}{\pi N} \sin(k_N z_i^c) \sin(k_N \frac{D}{2})$$

For out phase feed

$$\tilde{J}_{iM} = \frac{4}{\Psi_p} \cos(\overline{M} \varphi_i^c) \cos(\beta s \frac{\Psi_a}{2}) \sin(\overline{M} \frac{\Psi_a}{2})$$

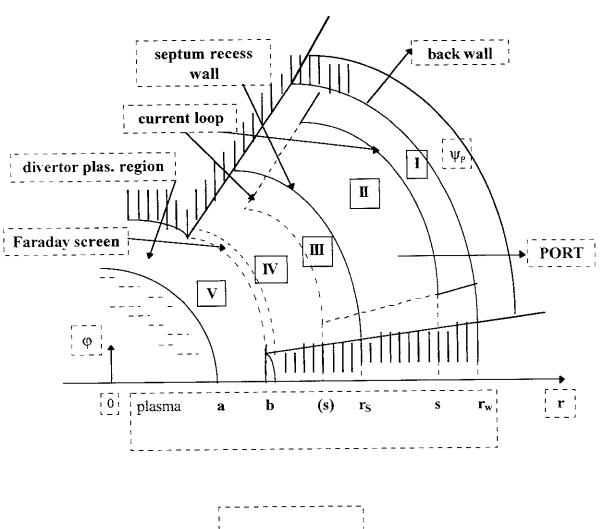
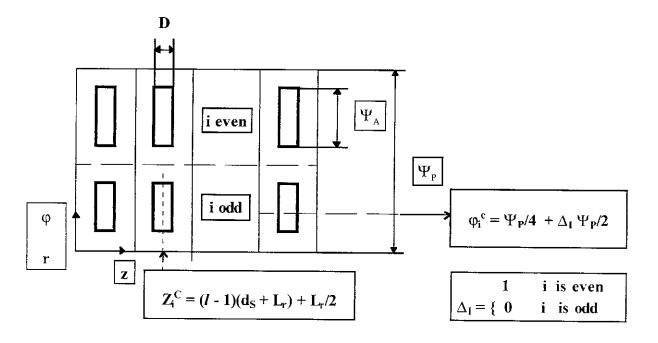


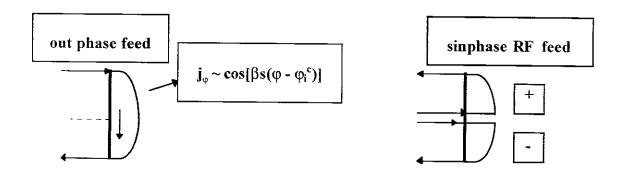
Fig. 1

Minor radius plasma cross section together with in-port multi loop ICRF antenna. In and out of recess loop location options are shown.

Fig.1ap. FOURIER COMPONENTS OF SUB LOOP CURRENTS



RF FEED of SUB LOOP



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